

Figure 1. Physical Photo of TEC18V15AD

### **FEATURES**

- The World's First TEC Controller Module with a Built-in Auto PID Compensation Network
- Input Voltage Range: 5.5V ~ 18V
- Output Voltage Range: –14.4V ~ 14.4V

 $@V_{VPS} = 18V$ 

- Output Current Range: 0 ~ 15A
- High Efficiency: >92%

 $@V_{VPS} = 18V \& V_{TEC} = 8V \& I_{TEC} = 15A$ 

- High Temperature Stability: <±0.001°C</li>
- Reduced Thermistor Injection Current: < 1µA</li>
- Continuous Bi-directional Output
- Programmable Output Current and Voltage Limits
- TEC Temperature, Current and Voltage Signals
- Selectable Temperature Sensor Types: thermistor, RTD, or temperature sensor IC
- High Reliability and Zero EMI
- Compact Size: 35.96 × 35.96 × 8.2 (mm)
- 100 % lead (Pb)-free and RoHS compliant

#### **APPLICATIONS**

Driving high power TEC modules at high efficiency without tuning the compensation network manually.

#### DESCRIPTION

TEC (Thermo-Electric Cooler) is a semiconductor device which can cool down or heat up the temperature of an object by injecting an electrical current in one or the other direction. This TEC controller, TEC18V15A, is designed to drive a TEC at

high efficiency for regulating the object temperature precisely by controlling the direction and magnitude of the current going through the TEC. It is powered by a DC voltage between 5.5V to 18V and the output current can go up to 15A without using a heat sink. Figure 1 is the photo of the actual controller TEC18V15AD.

The controller TEC18V15A allows setting the set-point temperature, maximum output voltage magnitude, and the maximum output current magnitude respectively. These three settings are the input parameters for the three control loops: constant temperature, constant current, and constant voltage. Before hitting the maximum output voltage magnitude or the maximum output current magnitude, the temperature loop is in control. When hitting the maximum output voltage magnitude, either outputting a positive or negative value across the TEC, the voltage loop takes over the control, the controller will be outputting a constant voltage to the TEC; when hitting the maximum output current magnitude, the current loop takes over the control, the controller will be outputting a constant output current to the TEC. The highest output voltage magnitude is limited by the maximum power supply voltage, and the maximum output current magnitude is 15A.

The temperature signal can be obtained by using one of these 3 temperature sensors: thermistor, RTD or temperature sensor IC. When using a thermistor, the set-point temperature range is determined by an external temperature network formed by 3 resistors. In order to reduce the injection current to the thermistor to reduce the errors caused by the self-heating effect, the injection current is provided in pulse mode, reducing the current by 10 times as opposed to a continuous current.

One advanced feature of this TEC controller is that it comes with a smart auto PID control micro-processor which senses and compensates for the thermal load automatically at real time, without needing external components to form a compensation network, nor requiring to fine tune the network with load tediously. The TEC controller with auto PID has a part number: TEC18V15ADAPID for the DIP package & TEC18V15ASAPID for the SMT package.

Conservative users can still choose using the conventional analog compensation network. The same as in the past, it requires a one-time pre-tuning the network to match the thermal load, but provides reliable and high accuracy control. For fixed thermal load applications, conventional analog compensation might be a good choice, while for applications with variable or numerous types of thermal loads, the

automatic PID control is more suitable.

Figure 2 is the top view of the controller, showing the pin names and the locations. There are a total of 32 pins in 2mm pitch. All the pins on the left are for either control input or indication output signals; all the right pins are power input or output.

The pin function details are given in Table 1.

At the thermistor input, there is a linearization circuit for the thermistor, to make the temperature output voltage be more linearly proportional to the actual thermistor temperature. There is a voltage inverter circuit, and it makes the temperature output voltage be positively proportional to the temperature, since the thermistor has a negative temperature coefficient. These 2 circuits together are called temperature measurement circuit. See Figure 6.

The set-point temperature voltage and the voltage representing the actual temperature are sent to an error amplifier. There is a compensation network inserted in the loop, to stop the oscillation of the controller caused by phase delay effects of the thermal load. Therefore, the compensation network must match the need for driving a particular thermal load. To simplify the tuning, a tunable compensation network is provided by the evaluation board for this TEC controller. A detailed guidance about how to tune

the compensation network with a thermal load is given in the evaluation board application note.

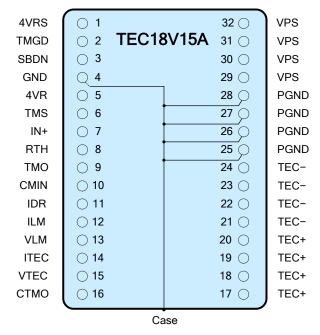


Figure 2. Pin Names and Location

#### **SPECIFICATIONS**

Table 1. Pin Function Descriptions

Pin #	Name	Note	Description
	4VRS	Analog output	4VR switch output. This pin outputs a switching pulse 4VR signal, from 0V to 4V, 85Hz, as a reference for the thermistor.
SNCO Digital output to 5V, 600kHz. It can be sent to the synchronization input (Switch Mode) controller or power supply, to eliminate the		Synchronization output. This pin outputs a switching pulse signal, from 0V to 5V, 600kHz. It can be sent to the synchronization input of another SM (Switch Mode) controller or power supply, to eliminate the beating interference between this TEC controller and the other SM device.	
2	TMGD	Digital output	Temperature good indication. Active high. Indicates when actual temperature equals to the set-point temperature of the target object. That is, the target object temperature is within $0.001^{\circ}\text{C}$ away from the set-point temperature, provided the set-point temperature range is $40^{\circ}\text{C}$ . Or $ V_{\text{TMO}}-V_{\text{TMS}}  < 0.5\text{mV}$ .  For the controller with PID function:  When the voltage of Pin 16 CTMO is 4V: when TMGD indicator is flashing, it indicates that it is identifying PID coefficient, and when this indicator is always on, the PID coefficient has been identified.

Pin #	Name	Note	Description
3	SBDN	Analog /Digital input	Standby and shut down control. This SBDN pin is internally floated and series with 1k resistor. It's suggested to pull this pin up to VPS power supply by a 4.99M $\Omega$ resistor. If pulled to ground, it shuts down the entire controller. This pin has 2 threshold voltages: 1.5V and 2.0V. See Figure 5. SHUT DOWN: V <sub>SBDN</sub> < 0.3V, the controller is set to non-working state. STANDBY: 1.9V > V <sub>SBDN</sub> > 1.5V, all components are set to working state except the output stages for TEC+ and TEC OPERATION: V <sub>SBDN</sub> > 2.0V, the whole controller is set to working state.
4	GND	Ground	Signal ground. Connect this pin to the signal ground of ADCs, DACs, and the signal sources. It can also be used as analog output pin ground.
5	4VR	Analog output	Reference voltage output, 4.096V. It can be used as the voltage reference by the potentiometers or DACs for setting the analog ports, such as TMS, ILM, VLM, etc. It can also be used by ADCs for sensing the analog output ports: TMO, CTMO, ITEC and VTEC. The initial accuracy is 0.1%, and the temperature coefficient is <50ppm/°C max.
6	TMS	Analog input	Analog Input port for setting the set-point temperature for the target object. It is internally tied a $1M\Omega$ resistor to the half value of the reference voltage, 2V. The open circuit voltage of this pin is thus 2V, corresponding to a set-point temperature of 25°C by using the default temperature network (with the set-point temperature range being from 15°C to 35°C). It is highly recommended to set this pin's voltage by using the controller's 4V voltage reference. This pin can be set by using a POT or DAC. When the set-point temperature needs to be at 25°C, leave this pin unconnected. Note: The max. voltage range of TMS pin is 0V-4V. It's recommended to set the voltage of this pin to 0.2V-3.8V, to prevent the controller from working in the saturation region.
7	IN+	Analog input	Receive external temperature signal (thermistor and temperature sensor, etc.)
8	RTH	Analog input	Thermistor connection port. Connect to the thermistor which is mounted on the target object for sensing its temperature. By using the default internal temperature network, a $10k\Omega$ @ $25^{\circ}$ C thermistor can be used. Other type of thermistors or temperature sensors can also be used, see the application section for details.
9	тмо	Actual target object temperature indication. It swings from 0V to 4V. It using a default internal temperature network, it represents 15°C to 35° output when this pin's voltage swings 0.1V to 3.9V linearly, provided a standa $10k\Omega$ thermistor is used as the temperature sensor device.	
10	CMIN	Analog input	Compensation input pin for the thermal control loop.
11	IDR	Analog input and output	This voltage is derived from the temperature error detection circuit and used as the input control signal of the current loop for the TEC. Its internal impedance is $10k\Omega$ and can be over-driven by an external analog signal which is able to over-ride the $10k\Omega$ resistor. The voltage range is from 0V to 4V, corresponding to $-15A$ to $+15A$ output current. Setting this pin voltage to 2V forces the output current to zero.



Pin #	Name	Note	Description
12	ILM	Analog input	This pin sets the TEC Current Limit. The maximum limit current is 15A. Setting this pin's voltage from 0V to 4V corresponds to setting the current magnitude limit from 0A to 15A.
13	VLM	Analog input	This pin sets the TEC voltage Limit. The maximum limit voltage is 30V. Setting this pin's voltage from 0V to 4V corresponds to the TEC voltage magnitude limit being from 0 to 30V: $V_{VLM} = \frac{ V_{TEC+} - V_{TEC-} _{MAX}}{7.5}$
14	ITEC	Analog output	TEC current indication. ITEC is an analog voltage output pin with a voltage proportional to the actual current through the TEC. ITEC's center voltage is 2V, corresponding to zero current through the TEC. $V_{\text{ITEC}} = \frac{I_{\text{OUT}}(A)}{7.5} + 2V, \text{ where } I_{\text{OUT}} \text{ is the actual output current of the controller,} $ flowing out from TEC+ port and flowing in to TEC- pin.
15	VTEC	Analog output	TEC voltage indication. VTEC is an analog voltage output pin with a voltage proportional to the actual voltage across the TEC. It swings from 0V to 4V to indicate the output voltage being from $-30V$ to $30V$ , so the center voltage is $2V$ . $V_{VTEC} = \frac{V_{TEC+} - V_{TEC-}}{15} + 2V$
16	стмо	Analog output	The controller internal temperature indication output. It can be used for sensing the actual temperature of the controller, to avoid over-heating. 0V to 4V sets the internal temperature from -55°C to 125°C linearly.  For the controller with PID function:  When there is a new load and the PID coefficient needs to be re-identified, the voltage of the pin is 4V.
17, 18, 19, 20	TEC+	Analog power output	This pin is for connecting to the positive terminal of the TEC module, all 4 pins are internally connected for increasing the current capability.
21, 22, 23, 24	TEC-	Analog power output	This pin is for connecting to the negative terminal of the TEC module, all 4 pins are internally connected.
25, 26, 27, 28	PGND	Power ground	Power ground for connecting to the power supply 0V return node, all 4 pins are internally connected.
29, 30, 31, 32	VPS	Power input	Power supply voltage positive node. The normal operating voltage range is 5.5V to 18V, the maximum value is 18V. All 4 pins are internally connected.

<sup>\*</sup>There are two part numbers for selection, TEC18V15A and TEC18V15ASNCO. The former's pin 1 is 4VRS, and the latter's pin 1 is SNCO. It's recommended to use TEC18V15A.

Table 2. Electrical characteristics.

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Units
Reference Voltage Pulse Out	put Mode: 4	VRS pin (Or Synchroniza	tion Out	put: SN	CO pin),	pin 1
Output Voltage Range	V <sub>4</sub> VRSOUT	T <sub>A</sub> = 25°C	4.0925	4.096	4.0995	V
Initial Error	VE	T <sub>A</sub> = 25°C	-0.05		0.05	%
Temperature Coefficient	T <sub>C</sub>			±3	±8	ppm/°C
Maximum Load Current	I <sub>4</sub> VRMAX	T <sub>A</sub> = 25°C	-20		+20	mA
Switch frequency	F <sub>4VRS</sub>		83	85	87	Hz
Output Voltage (Open circuit)	V <sub>SNCOOUT</sub>	Open circuit voltage = 0V ~ 4V PWM	0		4	V
Voltage Range (with load)	V <sub>SNCOOUT</sub>	Open circuit voltage = 0V ~ 4V PWM	0.1		3.9	V
Frequency	F <sub>SNCO</sub>	Open circuit voltage = 0V ~ 4V PWM		600		kHz
Temperature Good Indication	n: TMGD pin	, pin 2				
Voltage Range (Open circuit)	VTMGDOUT	Open circuit voltage = 4V	0		4	V
Voltage Range (with load)	V <sub>TMGDOUT</sub>	Open circuit voltage = 4V	0		4	V
Maximum Sourcing Current	ITMGDSC	Open circuit voltage = 4V	1		15	mA
Maximum Sourcing Voltage	V <sub>TMGDSC</sub>	Open circuit voltage = 4V	3.7		4	V
Maximum Sinking Current	ITMGDSK	Open circuit voltage = 4V	3		20	mA
Maximum Sinking Voltage	V <sub>TMGDSK</sub>	Open circuit voltage = 4V	0		0.6	V
Standby Shutdown Control:	SBDN pin, pi	n 3				
		$V_{SBDN} = 0V$	0.1		0.3	
Input Current	I <sub>SBDNIN</sub>	$V_{SBDN} = 4V$	4		6	μA
		V <sub>SBDN</sub> = 30V	30		50	
Input Voltage Range	V <sub>SBDNIN</sub>	Open circuit voltage = 5V	0		18	V
Shutdown Logic Low	VSBDNSDL	Open circuit voltage = 5V	0			V
Shutdown Logic High	V <sub>SBDNSDH</sub>	Open circuit voltage = 5V			0.7	V
Standby Logic Low	V <sub>SBDNSBL</sub>	Open circuit voltage = 5V	1.4			V
Standby Logic High	V <sub>SBDNSBH</sub>	Open circuit voltage = 5V			1.9	V
Operation Logic Low	V <sub>SBDNOPL</sub>	Open circuit voltage = 5V	2.0			V
Operation Logic High	V <sub>SBDNOPH</sub>	Open circuit voltage = 5V			5	V
Reference Voltage Output: 4	VR pin, pin 5	;				
Output Voltage Range	V <sub>4</sub> VROUT	T <sub>A</sub> = 25°C	4.0925	4.096	4.0995	V
Initial Error	VE	T <sub>A</sub> = 25°C		0.05		%
Temperature Coefficient	Tc	T <sub>A</sub> = -40°C ~ 125°C		3	8	ppm/°C
Maximum Load Current	I <sub>4</sub> VRMAX	T <sub>A</sub> = 25°C	-20		+20	mA
Maximum Load Capacitance	C <sub>4</sub> VRMAX		0.1		1	uF



Parameter	Symbol	Conditions	Min.	Тур.	Max.	Units
Temperature Set: TMS pin, pin	6			•	<u> </u>	
Input Impedance (See Figure 3 in Page 8 for input equivalent circuit)	Z <sub>TMSIN</sub>			5		ΜΩ
Input Voltage Range	$V_{TMSIN}$		0		4	V
Open Circuit Voltage	$V_{TMSOP}$			2		V
Temperature Signal Input: IN-	pin, pin 7					
Input Voltage Range	$V_{\text{IN+}}$		0		4	V
Thermistor Connection Port: R	TH pin, pin	8				
Input Voltage Range	V <sub>RTHIN</sub>		0		4	V
Actual Target Object Temperat	ure Indicat	ion: TMO pin, pin 9			<u> </u>	
Output Voltage Range	V <sub>TMOOUT</sub>	$R_{LOAD} = 10k\Omega$ to 2V -40°C≤ $T_A$ ≤ +125°C	0		4	V
Output Current	Ітмооит	$V_{SS} = 0V$ $T_A = 25^{\circ}C$	-12		12	mA
Compensation Input: CMIN pir	n, pin 10					
Input Voltage Range	V <sub>CMIN</sub>	$R_{LOAD} = 10k\Omega$ to 2V -40°C≤ $T_A$ ≤ +125°C	0		4	V
Input Current	I <sub>CMIN</sub>	-40°C≤ T <sub>A</sub> ≤ +125°C		90	200	pА
Compensation Output: IDR pin	, pin 11					
Output Voltage Range	VIDROUT	$R_{LOAD} = 10k\Omega$ to 2V -40°C≤ $T_A$ ≤ +125°C	0		4	V
TEC Current Limit: ILM pin, pin	12					
Input Impedance	Z <sub>ILM</sub>			21		kΩ
Input Voltage Range	VILMIN		0		4	V
TEC Voltage Limit: VLM pin, pir	າ 13	_	_	-		
Input Impedance (See Figure 4 in Page 8 for input equivalent circuit)	Z <sub>VLM</sub>			10		kΩ
Input Voltage Range	$V_{VLMIN}$		0		4	V
TEC Current Indication: ITEC p TEC Voltage Indication: VTEC p Controller Temperature Indica	oin, pin 15	pin, pin 16				
Output Voltage Range	V <sub>СТМО</sub>	$T_A = 25^{\circ}C$	0		4	V
Maximum Load Current	$I_{CTMOOUT}$	T <sub>A</sub> = 25°C	-12		12	mA
TEC+/TEC- pin, pin 17~20/pi	n 21~24					
Maximum Output Current	Imaxtec+   Imaxtec-	$V_{VPS} = 5.5V \sim 18V$ $T_A = 25^{\circ}C$	0		15	А
Maximum Output Voltage	Voutmax	$V_{VPS} = 18V$	0		14.4	V

Parameter	Symbol	Conditions	Min.	Тур.	Max.	Units
Power Supply Input: VPS pin, pin 29~32						
Input Voltage Range	V <sub>VPS</sub>		5.5		18	V
	I <sub>VPS</sub>	Operation mode	0.05		16	Α
Input Current	I <sub>VPSSB</sub>	Standby mode	5		20	mA
	I <sub>VPSSD</sub>	Shutdown mode			50	μΑ
Temperature Stability	Temperature Stability					
Temperature Error Voltage	V <sub>TMO</sub> -V <sub>TMS</sub>		-0.47	0.02	0.47	mV
Efficiency	η	V <sub>VPS</sub> =18V  V <sub>TEC+</sub> - V <sub>TEC-</sub>  =14V  I <sub>TEC+</sub> - I <sub>TEC-</sub>  =15A		≥92		%
Case Operating Temperature Range	Tcs		-40		110	°C
Ambient Operating Temperature Range	T <sub>A</sub>		-40		65	°C
Storage Temp. Range	T <sub>STG</sub>		-40		125	°C
Controller Case Thermal Resistance	R <sub>TH</sub>			9		°C /W

This TEC controller can only drive the TECs having  $> 1\Omega$  impedance, which equals  $V_{MAX}/I_{MAX}$ .

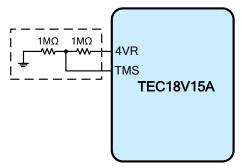


Figure 3. TMS Input Equivalent Circuit

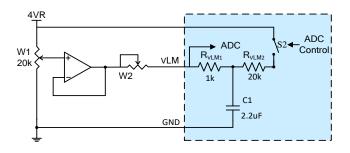


Figure 4. VLM Input Equivalent Circuit

The switch S2 is closed @ heating, and open @ cooling

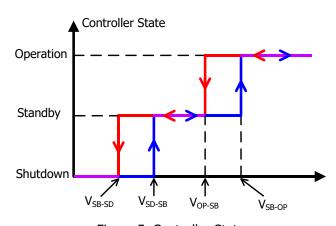


Figure 5. Controller States

 $V_{\text{SB-SD}}$ : Going down logic low from standby to shutdown

 $V_{\text{SD-SB}}$ : Going up logic high from shutdown to standby

V<sub>OP-SB</sub>: Going down logic low from operation to standby

V<sub>SB-OP</sub>: Going up logic high from standby to operation

## **BLOCK DIAGRAM**

The block diagram of the controller is shown in Figure 6.

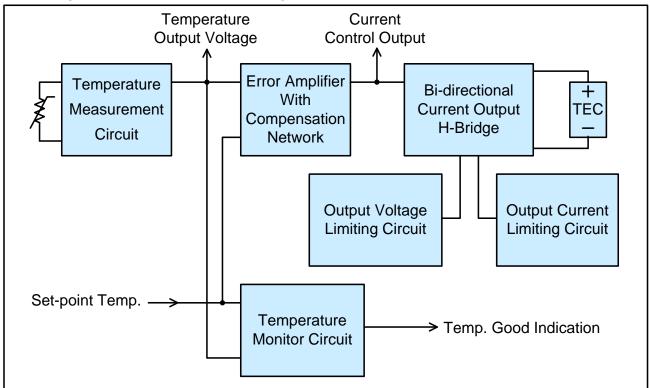


Figure 6. TEC Controller Block Diagram

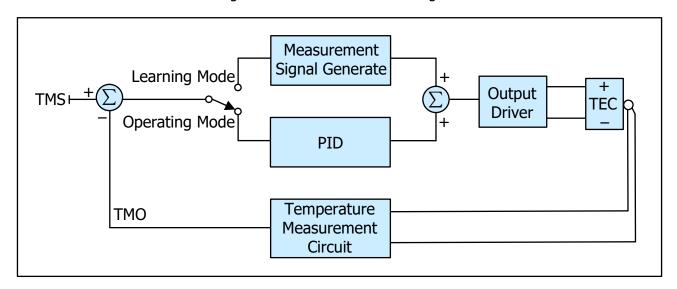


Figure 7. Auto PID System Block Diagram

## **APPLICATIONS**

TEC controller connections are shown in Figure 8.

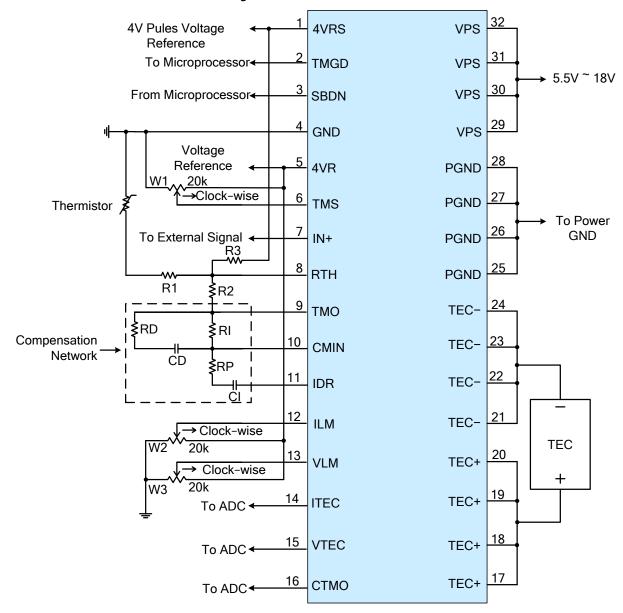


Figure 8.1. TEC Controller Connection

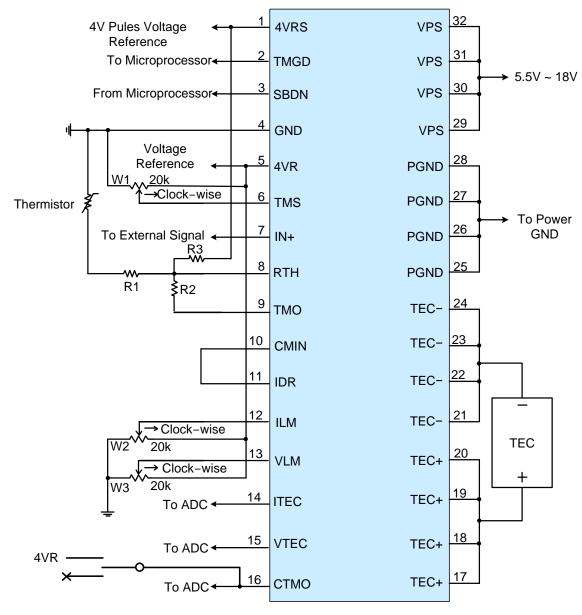


Figure 8.2. Connection for TEC18V15ADAPID &TEC18V15ASAPID

#### **SBDN**

Table 3. External Detector Selection.

No.	Input	Voltage	Controller State	Temperature Sensor Type and Mode
1		≤ 0.5V	Shut Down	-
2		1.7V ± 0.2V	Standby	-
3	SBDN	$2.1V \pm 0.1V$		Temperature Sensor IC or RTD
4		2.4V ± 0.1V	Working	Thermistor continuous mode
5		2.9V ± 0.1V		Thermistor in pulse mode

#### For the controller with PID function:

No.	Input	Voltage	Controller State	Temperature Sensor Type and Mode	PID Type
1		≤ 0.5V	Shut Down	-	-
2		1.5V ± 0.2V	Standby	-	-
3		2.1V ± 0.1V		Temperature sensor IC or RTD	
4	SBDN	$2.4V \pm 0.1V$		Thermistor continuous mode	External Analog PID
5		2.9V ± 0.1V	Working	Thermistor pulse mode	7 11 1010 9 1 12
6		$3.2V \pm 0.1V$	_	Temperature sensor IC or RTD	
7		$3.6V \pm 0.1V$		Thermistor continuous mode	Auto PID
8		$3.9V \pm 0.1V$		Thermistor pulse mode	

#### **Temperature Sensor Selections**

There are usually three temperature sensors, thermistor, RTD (Resistance Temperature Detector), and IC (Integrated Circuit) temperature sensors.

#### 1. Thermistor

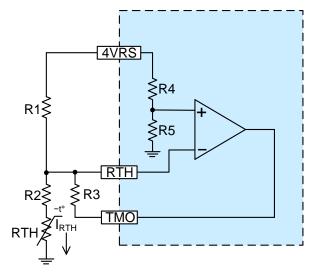
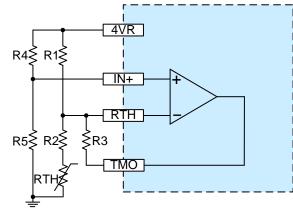


Figure 9.1. RTH (Pulse Mode)



Note: R4=R5

Figure 9.2. RTH

To achieve the required  $V_{\text{TMO}}$  outputs at the three different setting point temperatures in the Temperature Network, use the equation:

Temperature Network, use the equation: 
$$R1 = R_{MID} + \frac{R_{MID} \times \left(R_{LOW} + R_{HIGH}\right) - 2 \times R_{HIGH} \times R_{LOW}}{R_{HIGH} + R_{LOW} - 2 \times R_{MID}} \quad \textbf{(1)}$$

$$R2 = R1 - R_{MID} (2)$$

$$R3 = \frac{R1 \times \left(R1 + R_{LOW} - R_{MID}\right)}{R_{LOW} - R_{MID}} \quad \textbf{(3)}$$

For example, setting the high set-point temperature at  $35^{\circ}$ C and the low set-point temperature at  $15^{\circ}$ C results in a middle set-point temperature  $(35 + 15)/2 = 25^{\circ}$ C. Use the R-T table of a thermistor.

$$R_{HIGH} = 6.9k\Omega$$

 $R_{MID} = 10k\Omega$ 

 $R_{IOW} = 14.8 k\Omega$ 

Note that Equation 1 to Equation 3 result in

$$R1 = 17.5k\Omega$$

 $R2 = 7.5k\Omega$ 

 $R3 = 81.3k\Omega$ 

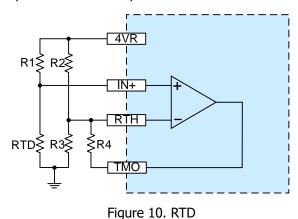
In order to reduce the injection current to the thermistor to reduce the errors caused by the self-heating effect, the injection current is provided in pulse mode, reducing the current by 10 times as opposed to a continuous current.

It's recommended to connect R1 to 4VRS, and the controller will measure temperature at intervals that will reduce the error caused by the RTH self-heating. At the same time, the SBDN pin should be between 3.1V and 4V. See Table 3.

We can also connect R1 to 4VR, but it may lead to some errors caused by RTH self-heating. At the same time, SBDN pin should be between 2.4V and 2.6V. See Table 3.

#### 2. RTD

RTD is short for resistance temperature detector, which features high accuracy and low drift. It usually generates heat when the current flows through the RTD, which is called self-heating effect. Moreover, RTD has an approximate linear resistance-temperature relationship.



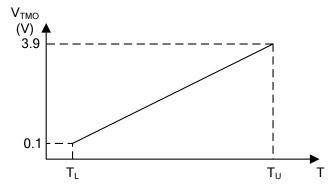


Figure 11. Linear Relationship between V<sub>TMO</sub> and Temperature

$$R_{TD} = R_0 \times (1 + 0.00385T)$$

e.g. 
$$R_0 = 1k\Omega$$

When T = 10°C, 
$$R_{TD}(10) = 1.0385$$
kΩ

When T = 40°C, 
$$R_{TD}(40) = 1.154k\Omega$$

Choose R1

A.  $P_{RTD} \le 1$  mW,  $R_{TD} = 1000\Omega$ 

$$P_{RTD} = (I_{RTD})^2 \times 1000\Omega = 0.001W$$

$$I_{RTD} = 1mA = \frac{4VR}{R1 + R_{TD}} = \frac{4}{R1 + 1k} \longrightarrow R1 = 3k\Omega$$

B.  $P_{RTD} \le 1$  mW,  $R_{TD} = 100\Omega$ 

$$P_{RTD} = (I_{RTD})^2 \times 100\Omega = 0.001W$$

$$I_{RTD} = 3.16 \text{mA} = \frac{4 \text{VR}}{R1 + R_{TD}} = \frac{4}{R1 + 0.1 \text{k}}$$

 $R1=1.15k\Omega$ 

$$V_{TMO} = \frac{4 \times R_{TD}}{R1 + R_{TD}} \times \left[ 1 + \frac{R4 \times (R2 + R3)}{R2 \times R3} \right] - \frac{4 \times R4}{R2}$$

I. When T = 10°C, R1 = 3kΩ,  $R_{TD}(T_L) = 1.0385kΩ$ ,

$$0.93 = \frac{R4 \times (2.97R3 - 1.03R2)}{R2 \times R3}$$

When T = 40°C,  $R1 = 3k\Omega$ ,  $R_{TD}(T_U) = 1.154k\Omega$ ,

$$2.79 = \frac{R4 \times (1.11R2 - 2.89R3)}{R2 \times R3}$$

II. When  $T = 10^{\circ}C$ ,  $R1 = 1.15k\Omega$ ,  $R_{TD}(T_L) = 1.0385k\Omega$ ,

$$1.8 = \frac{R4 \times (2.1R3 - 1.9R2)}{R2 \times R3}$$

When  $T = 40^{\circ}C$ ,  $R1 = 1.15k\Omega$ ,  $R_{TD}(T_U) = 1.154k\Omega$ ,

$$1.9 = \frac{2 \times R4 \times (R2 - R3)}{R2 \times R3}$$

To achieve the required  $V_{\text{TMO}}$  outputs at the three different setting point temperatures in the Temperature Network, use the equation:

When T = LOW, RTD = RTD<sub>L</sub>, TMO = 
$$0.1V$$
,  $V_1 = V_{1L}$ 

When T = HIGH, RTD = RTD<sub>H</sub>, TMO = 
$$4.0V$$
,  $V_1 = V_{1H}$ 

$$\Delta TMO = 4V - 0.1V = 3.9V$$

$$\Delta V1 = V_{1H} - V_{1L}$$

$$G \,=\, \frac{\Delta TMO}{\Delta V1} \,=\, 1 \,+\, \frac{R4 \,\times\, (R2 \,+\, R3)}{R2 \,\times\, R3}$$

$$RTD = R_0 \times (1+0.00385T)$$

e.g. 
$$R_0 = 1k\Omega$$

$$V1 \,=\, 4.096V \times \frac{RTD}{R1 + RTD}$$

$$V_{1L} = 0.5V$$

$$R1 = RTD_L \times \frac{4.096V}{V_{11}} - RTD_L$$

$$R2 = R1$$
,  $R3 = RTD_L$ 

$$R4 = (G - 1) \times \frac{R2 \times R3}{R2 + R3}$$

For example, setting the high set-point temperature at 60°C and the low set-point temperature at 0°C.Use RTD =  $R_0 \times (1+0.00385T)$ , (e.g.  $R_0 = 1k\Omega$ ).

$$RTD_L = R_{TD}(10^{\circ}C) = 1.0k\Omega$$

$$RTD_H = R_{TD}(60^{\circ}C) = 1.231k\Omega$$

$$R_1 = 7.192k\Omega$$

$$R_2 = R_1 = 7.192k\Omega$$

$$R_3 = RTD_L = 1.0k\Omega$$

$$R_4 = 32.308k\Omega$$

3. IC

#### (1). Positive Coefficient

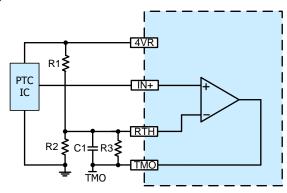


Figure 12.1. PTC IC temperature sensor

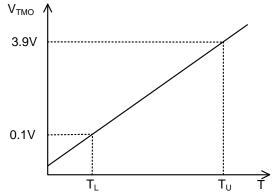


Figure 12.2. Objective

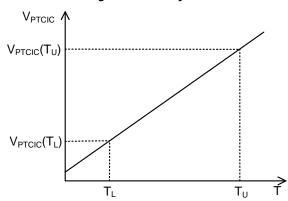


Figure 12.3. Temperature vs. V<sub>PTCIC</sub>

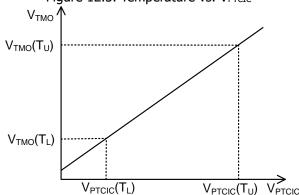


Figure 12.4. VPTCIC VS. VTMO

$$V_{TMO}(T_L) = 0.1V, V_{TMO}(T_U) = 3.9V$$

$$G \,=\, \frac{\Delta V_{\text{TMO}}}{\Delta V_{\text{PTCIC}}} \,=\, \frac{V_{\text{TMO}}(T_{\text{U}}) \,-\, V_{\text{TMO}}(T_{\text{L}})}{V_{\text{PTCIC}}(T_{\text{U}}) \,-\, V_{\text{PTCIC}}(T_{\text{L}})}$$

$$G\,=\,\frac{R3}{R1/\!/R2}\,+\,1$$

$$V_{PTCIC}(T_{M}) \; = \; \frac{V_{PTCIC}(T_{U}) \; + \; V_{PTCIC}(T_{L})}{2}$$

$$V_{PTCIC} = V_{PTCIC}(T_{M}), V_{TMO} = \frac{3.9V + 0.1V}{2} = 2V$$

$$\frac{V_{PTCIC}(T_{M})}{R2} = \frac{2V - V_{PTCIC}(T_{M})}{R3} + \frac{4V - V_{PTCIC}(T_{M})}{R1}$$

 $R3=20k\Omega$ 

$$R2 \, = \, \frac{R3}{[4V \, - \, V_{\text{PTCIC}}(T_{\!_{M}})] \times G \, - \, 2} \label{eq:R2}$$

$$R1 = \frac{R2 \times R3}{R2 \times (G-1) - R3}$$

#### 2. Negative Coefficient

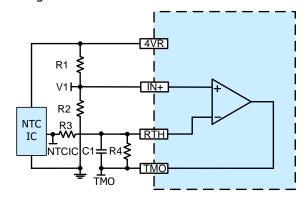


Figure 13.1. NTC IC temperature sensor

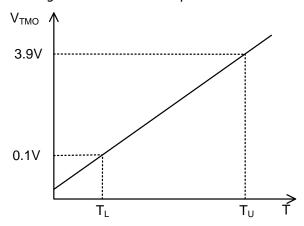


Figure 13.2. Objective

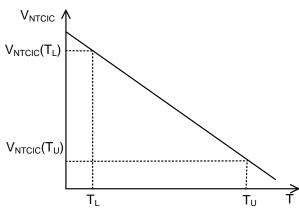


Figure 13.3. Temperature vs. V<sub>NTCIC</sub>

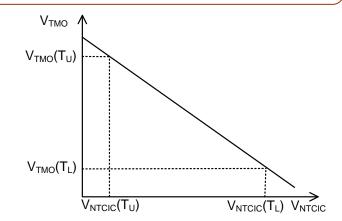


Figure 13.4. VPTCIC vs. VTMO

$$V_{TMO}(T_L) = 0.1V, V_{TMO}(T_U) = 3.9V$$

$$G \,=\, \frac{\Delta V_{\text{TMO}}}{\Delta V_{\text{NTCIC}}} \,=\, \frac{V_{\text{TMO}}(T_{\text{U}}) \,-\, V_{\text{TMO}}(T_{\text{L}})}{V_{\text{NTCIC}}(T_{\text{L}}) \,-\, V_{\text{NTCIC}}(T_{\text{L}})}$$

$$G = \frac{R4}{R3}$$

$$R4 = 20k\Omega \sim 200k\Omega$$

$$R3 = \frac{R4}{G}$$

$$V_{\text{NTCIC}}(T_{\text{M}}) = \frac{V_{\text{NTCIC}}(T_{\text{U}}) + V_{\text{NTCIC}}(T_{\text{L}})}{2}$$

$$\frac{[2V-V_{NTCIC}(T_M)]\times R3}{R3+R4}+V_{NTCIC}(T_M)\,=\,\frac{4V\times R2}{R1+R2}$$

R2=10k

$$R1 \, = \, \frac{40 \times (1 + G)}{2 - V_{NTCIC}(T_M)} - 10$$

$$\frac{1}{2\pi \times (R3//R4) \times C} = 200Hz$$

$$C = \frac{1}{2\pi \times (R3/\!/R4) \times 200Hz} = \frac{1 + G}{400\pi \times R4}$$

 $\label{eq:maximum sourcing current: } \frac{V_{\text{NTCIC}}(T_{\text{L}}) - V1}{\text{R3}} \, \leq \, I_{\text{SOURCEMAX}}$ 

Maximum sinking current:  $\frac{V1 - V_{\text{NTCIC}}(T_{\text{L}})}{R3} \leq I_{\text{SOURCEMAX}}$ 

#### **SBDN**

SBDN is suggested to be pulled up to VPS with a  $10\mu\text{A}$  current and contains a 1.50V logic threshold. Drive this pin to a logic-high to enable the TEC18V15A. Drive to a logic-low to disable the TEC controller and enter micro-power shutdown mode.

#### **ITEC and ILM**

When the voltage of the ITEC is  $V_{\text{ITEC}}=2V$ , the current of the TEC Controller  $I_{\text{TEC}}=0A$ . When  $V_{\text{ITEC}}=0V$ ,  $I_{\text{TEC}}$  has the maximum reverse current, -15A. When  $V_{\text{ITEC}}=4V$ ,  $I_{\text{TEC}}$  has the maximum forward current, 15A.

TEC controller is working on the cooling region, when it has forward current. On the opposite, it works on the heating region when reversing the current, as shown in Figure 14.

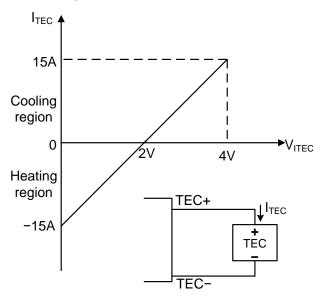


Figure 14. VITEC vs. ITEC

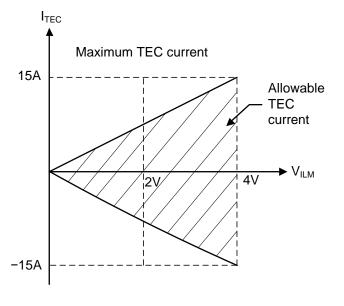


Figure 15. VILM vs. ITEC

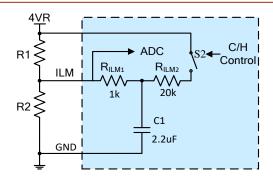


Figure 16. ILM vs. Cooling and Heating Control
The switch S1 is closed @ heating, and open @ cooling

Calculate the maximum current in cooling and heating region according to Figure 16.

1. Cooling region

I<sub>TEC</sub> $\ge 0A$ , V<sub>ILM</sub> $\ge 2V$ , Cooling region => S1 = Open;

Maximum cooling current:

$$I_{TEC} \le \frac{V_{ILM}}{4V} \times 15A = \frac{R2}{R1 + R2} \times 15A$$

2. Heating region

 $I_{\text{TEC}}$  < 0A,  $V_{\text{ILM}}$  < 2V, Heating region => S1 = Close;

Maximum heating current:

$$|I_{TEC}|_{MAX} \le \frac{V_{ILM}}{4V} \times 15A = \frac{R2//R_{ILM}}{R1 + R2//R_{ILM}} \times 15A$$

3. After deciding the heating current shrinking ratio, we can determine the value for R1 & R2.

Calculate R1 & R2 ratio

$$I_{COOLMAX} = \frac{R1}{R1 + R2} \times 15A$$
 -----(1)

Calculate R1 & R2 value by deciding the heating current shrinking ratio:

KHC = maximum heating current / maximum cooling current

$$= \frac{I_{\text{ITEC-(TH-MAX)}}}{I_{\text{ITEC-(CL-MAX)}}} -----(2)$$

$$= \frac{\frac{R2//R_{ILM}}{R1 + R2//R_{ILM}}}{\frac{R2}{R1 + R2}}$$

$$= \frac{200 \times (R1 + R2)}{R1 \times R2 + 200 \times (R1 + R2)}$$

#### **VTEC and VLM**

VTEC =  $V_{TEC+} - V_{TEC-}$ , as shown in Figure 18.

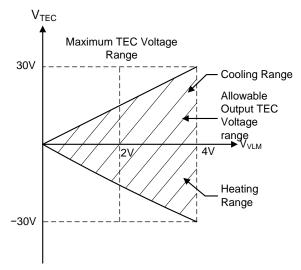


Figure 17. V<sub>TEC</sub> vs. V<sub>VLM</sub>

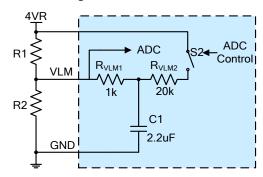


Figure 18. VLM vs. Cooling and Heating Control
The switch S2 is closed @ heating, and open @
cooling

#### **TMGD**

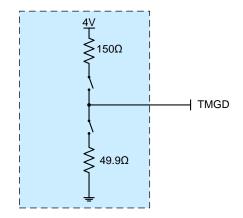


Figure 19. Output Equivalent Circuit for TMGD Pin

The TMGD pin outputs the maximum source current and sink current of 20mA. The output current will cause voltage drop, see Figure 19.

#### **VLM and ILM**

If you want to use this TEC controller for other applications not discussed here, such as with wave locker controllers, consult with us. The same for other customizations, such as setting the ILM and VLM by using voltage source swings above 4V and/or VPS.

An external voltage connects the ILM pin through a resistor. This voltage can be used to adjust the voltage range of cooling or heating, and advice is 1.5V. The resistor can be used to adjust the difference of cooling and heating, and advice is  $10k\Omega$ . See Figure 20.

For example, the voltage midpoint of the ILM pin  $(V_m)$  is 2V. Adjust the external voltage, and make the voltage range 1V, but it is only with the center of 2V  $(V_m)$ . If you adjust the resistor W2, you can move the limit of the cooling to be greater than the limit of the heating. It is shown in Figure 21 and Figure 22.

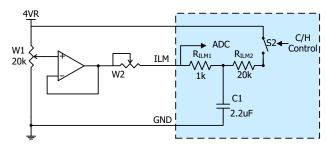


Figure 20. ILM vs. Cooling and Heating Control

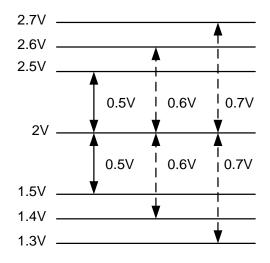


Figure 21. Adjust the External Voltage

We can tell the VLM or ILM voltage in cooling control or heating control through the waveforms on the VLM

or ILM pin, see Figure 23 and Figure 24. The duty cycle in Figure 23 is 99% and 1% in Figure 25. We can also measure both voltages by a multimeter. When the controller is in the Standby State, the voltage measured by the multimeter is the VLM or ILM voltage in cooling control. When the controller is in Operation State, the voltage measured by the multimeter is the VLM or ILM voltage in heating control.

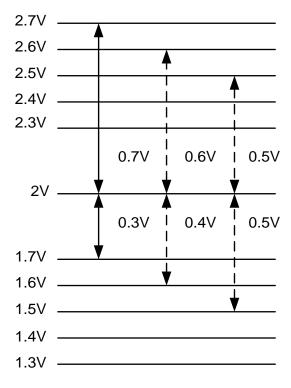


Figure 22. Adjust the Resistor

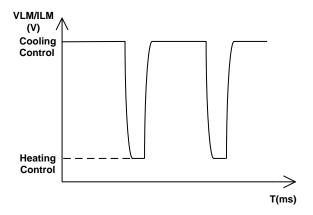


Figure 23. The Waveform on the VLM or ILM Pin

@ SB State

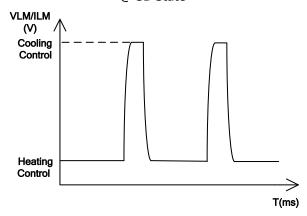


Figure 24. The Waveform on the VLM or ILM Pin

@ Operation State

#### **Temperature-Network**

TEC18V15A comes with a customized internal compensational network for which the component values are specified by the customer. See Figure 8.

TEC18V15A comes with a customized Temperature network. See Figure 6 and Figure 8.

## **TYPICAL CHARACTERISTICS**

Table 4. Measurement Data of Rth vs. Temperature

Temp. (°C)	Rth (kΩ)	TMO (V)	Ideal Linear (V)	Error (V)
15	15.7049	0.05	0.1	-0.05
16	14.9944	0.24	0.3	-0.06
17	14.3198	0.43	0.49	-0.06
18	13.6792	0.63	0.69	-0.06
19	13.0705	0.82	0.88	-0.06
20	12.4922	1.02	1.08	-0.06
21	11.9425	1.22	1.27	-0.05
22	11.4198	1.42	1.47	-0.05
23	10.9227	1.62	1.66	-0.04
24	10.4499	1.82	1.86	-0.04
25	10	2.03	2.05	-0.02

Temp. (°C)	Rth (kΩ)	TMO (V)	Ideal Linear (V)	Error
26	9.5718	2.23	2.25	-0.02
27	9.1642	2.44	2.44	0
28	8.776	2.64	2.64	0
29	8.4063	2.85	2.83	0.02
30	8.0541	3.05	3.03	0.02
31	7.7184	3.25	3.22	0.03
32	7.3985	3.46	3.42	0.04
33	7.0935	3.66	3.61	0.05
34	7.0935	3.86	3.81	0.05
35	6.5251	4.06	4.00	0.06

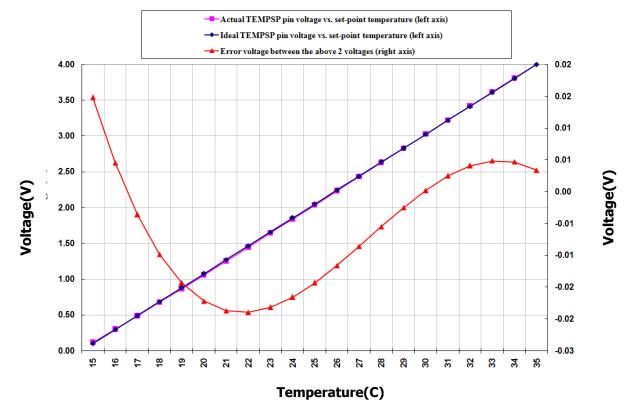


Figure 25. TMO Pin Voltage vs. Set-point Temperature

## **HEAT DISSIPATION**

The heating elements of the TEC18V15A TEC controller are on the top layer next to the shell, so the heat sink needs to be installed on the top of the controller shell. When the output current of the

controller is <8A, no heat sink is required. When the output current is >8A, a heat sink is needed to ensure the internal temperature of the controller is <85°C.

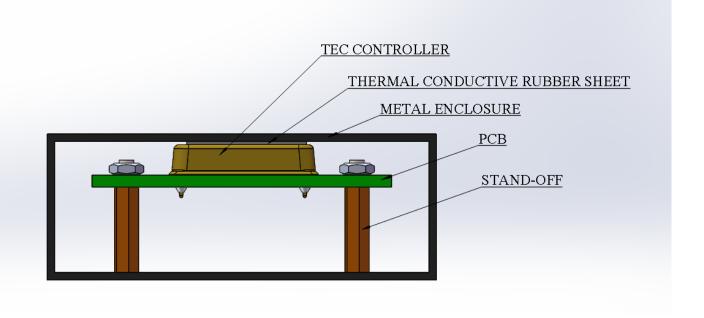


Figure 26. Transferring Heat with Metal Enclosure

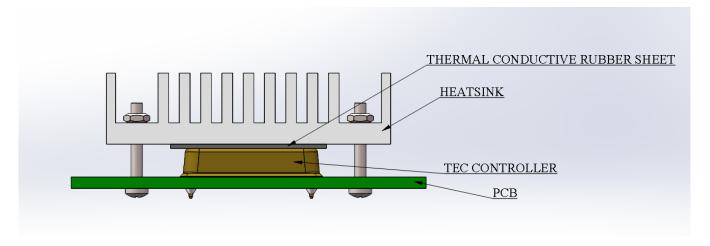
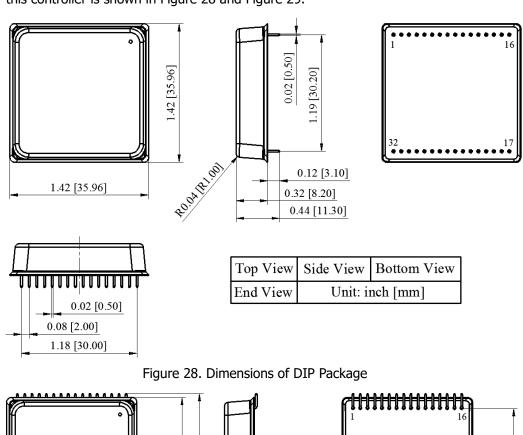


Figure 27. Transferring Heat with Heat Sink

## **MECHANICAL DIMENSIONS**

The controller comes in 2 packages: through-hole mount and surface mount. The former is often called DIP (Dual Inline package) or D (short for DIP) package and has a part number: TEC18V15AD, and the latter is often called SMT (Surface Mount Technology) or SMD (Surface Mount Device) package and has a part number: TEC18V15AS. Dimensions of this controller is shown in Figure 28 and Figure 29.



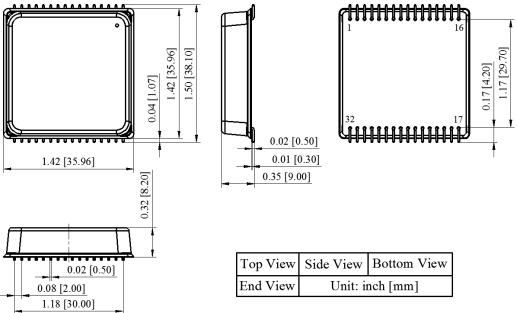


Figure 29. Dimensions of SMT Package

## **High Voltage High Current TEC Controller**

TEC18V15A

## **ORDERING INFORMATION**

Table 5. Part Number

Part Number	Description
TEC18V15AD	DIP package, with Pin 1 4VRS.
TEC18V15AS	SMT package, with Pin 1 4VRS.
TEC18V15ASNCOD	DIP package, with Pin 1 SNCO.
TEC18V15ASNCOS	SMT package, with Pin 1 SNCO.
TEC18V15ADAPID	DIP package, with Pin 1 4VRS and with auto PID function.
TEC18V15ASAPID	SMT package, with Pin 1 4VRS and with auto PID function.

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